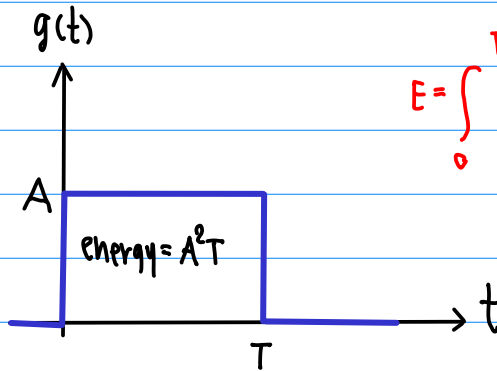
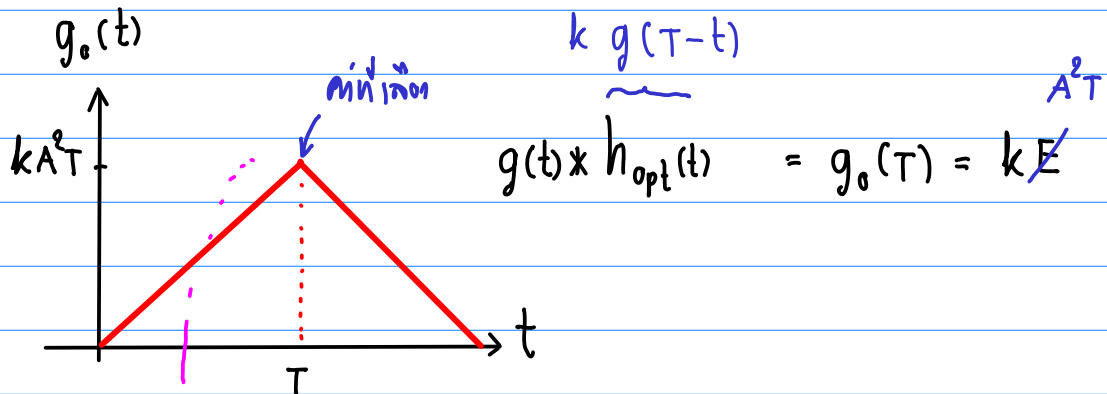


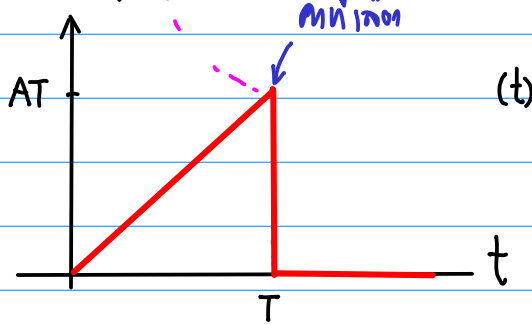
Ex *

$$E = \int_0^T |g(t)|^2 dt$$



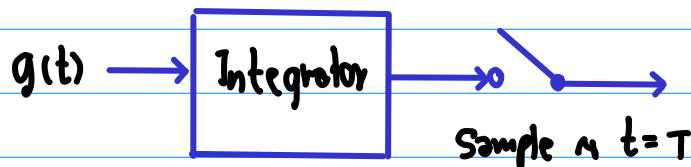
$$g(t) * h_{opt}(t) = g_o(t) = kA^2$$

o/p of integrate-and-dump circuit



$$g_u \quad kA = 1$$

สัญญาณอินพุต



4.3 Error Rate Due to Noise

ໃນ PCM system (binary) based on polar non-return-to-zero (NRZ) signaling

channel noise ເປັນ additive white Gaussian noise

ກັບ mean ເປັນ zero ແລະ power spectral density ໃຫຍ່ N_0

ເຄື່ອງໝາຍ ນີ້ ລຽນ ການ ດຳລົງ

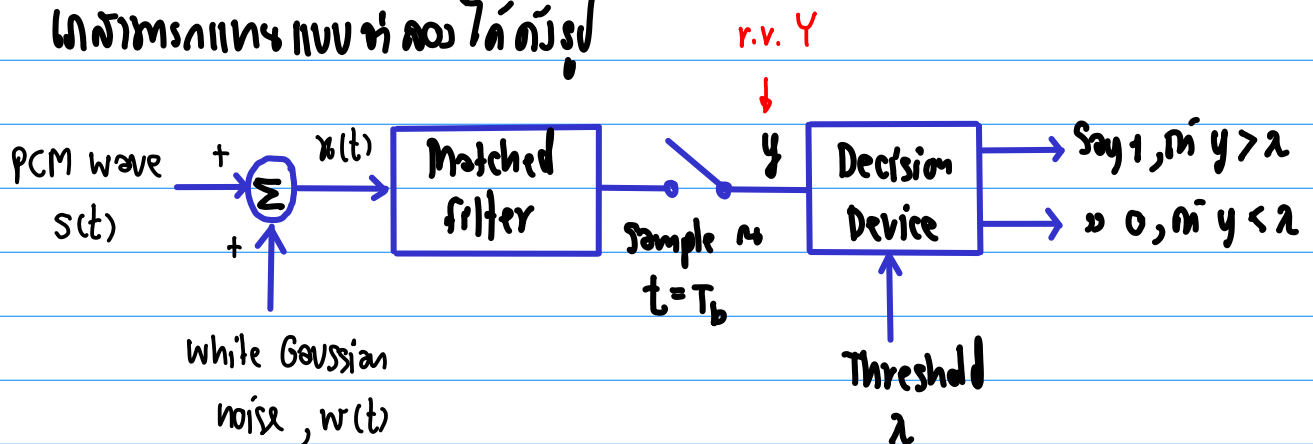


Fig: Receiver ສຳລັບ baseband transmission (polar NRZ)

ໃນ T_b ເປັນໄລຍະເວລາ ການສົ່ງ Pulse 1 bit

ໄດ້ received signal ດັ່ງນັ້ນ

$$x(t) = \begin{cases} +A + w(t), & \text{Symbol 1 ອຸກສັ່ງ} \\ -A + w(t), & \text{Symbol 0 ອຸກສັ່ງ} \end{cases} \quad (4.11)$$

ຈາກນີ້ ຈະໄດ້ ລິ້ງເກີດ ຂໍ້ ຜິດພາດ ເປັນ error 2 ກໍລະນີ

1. Symbol "0" ອຸກສັ່ງ ແຕ່ໄດ້ຮັບ decision ເປັນ "1" ເຊິ່ງເອີ້ນວ່າ **error of first kind**
2. Symbol "1" ອຸກສັ່ງ ແຕ່ໄດ້ຮັບ decision ເປັນ "0" ເຊິ່ງເອີ້ນວ່າ **error of second kind**

ปัญหาการ: error ให้มีพหุคูณ สัญลักษณ์ "0" หนึ่ง

ทบทวน (4.21) ให้น

$$x(t) = -A + w(t), \quad 0 \leq t \leq T_b \quad (4.22)$$

ทบทวน EX* ให้น $kAT_b = 1$ และ ให้น

$$\begin{aligned} y &= \int_0^{T_b} x(t) dt \\ &= -A + \frac{1}{T_b} \int_0^{T_b} w(t) dt \end{aligned} \quad (4.23)$$

ให้หาผลรวมของ ทบทวน Gaussian ให้น Y เป็น Gaussian $\left\{ \begin{array}{l} \text{mean} \checkmark \\ \text{variance} \checkmark \end{array} \right.$

mean:
$$\begin{aligned} E[Y] &= E \left[-A + \frac{1}{T_b} \int_0^{T_b} w(t) dt \right] \\ &= E[-A] + \frac{1}{T_b} \int_0^{T_b} E[w(t)] dt \\ &= -A \end{aligned}$$

variance:

$$\begin{aligned} \sigma_Y^2 &= E[(Y - \mu_Y)^2] = E[(Y + A)^2] \\ &= E \left\{ \left[\underbrace{\left(-A + \frac{1}{T_b} \int_0^{T_b} w(t) dt \right)}_Y + A \right]^2 \right\} \end{aligned}$$

(424)

$$= \frac{N_0}{2T_b} \quad (4.26)$$

150 "0" 99 53 104

$$f_Y(y|0) = \frac{1}{\sqrt{\pi N_0/T_b}} e^{-\left(\frac{(y+A)^2}{N_0/T_b}\right)} \quad (4.27)$$

၎င်း p_{10} ကို conditional probability of error for symbol "0" ခေါ်သည်

သို့

$$\begin{aligned}
 p_{10} &= P(y > \lambda \mid \text{symbol "0" ခေါ်သည်}) \\
 &= \int_{\lambda}^{\infty} f_Y(y|0) dy \\
 &= \frac{1}{\sqrt{\pi N_0/T_b}} \int_{\lambda}^{\infty} e^{-\left(\frac{(y+A)^2}{N_0/T_b}\right)} dy \quad (4.28)
 \end{aligned}$$

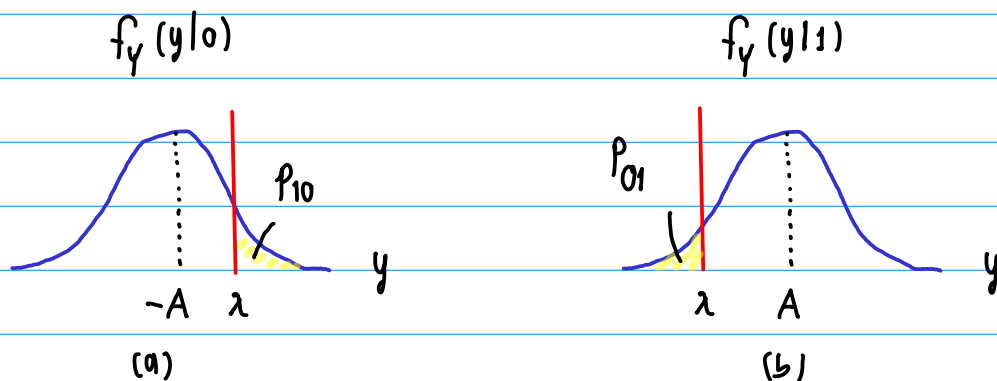


Fig. (a) pdf of r.v. Y as matched filter o/p with "0" ခေါ်သည်
(b) \rightarrow "1" ခေါ်သည်

သို့ Complementary error function (Gaussian distribution)

$$\text{erfc}(u) = \frac{2}{\sqrt{\pi}} \int_u^{\infty} \exp(-z^2) dz$$

ကို u ကို minimum ကို upper bound ကို Complementary error function ခေါ်သည်

$$\operatorname{erfc}(u) < \frac{\exp(-u^2)}{\sqrt{\pi} u} \quad (4.30)$$

ដូច្នេះ

$$z = \frac{y+A}{\sqrt{N_0/T_b}}$$

ដោយ ប្រើ (4.28) យើងបាន

$$\begin{aligned} p_{10} &= \frac{1}{\sqrt{\pi}} \int_{(A+\lambda)/\sqrt{N_0/T_b}}^{\infty} e^{-z^2} dz \\ &= \frac{1}{2} \operatorname{erfc}\left(\frac{A+\lambda}{\sqrt{N_0/T_b}}\right) \end{aligned} \quad (4.31)$$

ក្រាហ្វិកខាងលើនេះ ជួយឲ្យយើង "ប្រាប់" ពីលទ្ធភាព

$$f_Y(y|1) = \frac{1}{\sqrt{\pi N_0/T_b}} e^{-\left(\frac{(y-A)^2}{N_0/T_b}\right)} \quad (4.32)$$

យើង

$$\begin{aligned} p_{01} &= P(y < \lambda \mid \text{symbol "1" ត្រូវបានប្រើ}) \\ &= \int_{-\infty}^{\lambda} f_Y(y|1) dy \\ &= \frac{1}{\sqrt{\pi N_0/T_b}} \int_{-\infty}^{\lambda} e^{-\left(\frac{(y-A)^2}{N_0/T_b}\right)} dy \end{aligned} \quad (4.33)$$

$$q_u \quad z = \frac{A - y}{\sqrt{N \cdot T_b}}$$

11. (4.33) 0.104

$$p_{01} = \frac{1}{\sqrt{1}} \int_{(A-\lambda)/\sqrt{N_0/T_b}}^{\infty} e^{-z^2} dz$$

$$= \frac{1}{2} \operatorname{erfc} \left(\frac{A-\lambda}{\sqrt{N_0/T_b}} \right) \quad (4.79)$$

ใน p_0 และ p_1 นั้น priori probability ของ messages symbols "0"

115. "1" អាយត៍

ແຈ້

$$P_e = P_0 P_{10} + P_1 P_{01}$$

$$= \frac{P_0}{2} \operatorname{erfc}\left(\frac{A+\lambda}{\sqrt{N_0/T_b}}\right) + \frac{P_1}{2} \operatorname{erfc}\left(\frac{A-\lambda}{\sqrt{N_0/T_b}}\right) \quad (4.35)$$

Goal: minimize p_e